

Audio Signal Processing : IV. Stochastic signal processing

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Framework :

$X[n]$ is a stochastic discrete-time signal

$X[n]$ is real-valued

Stochastic signal : What for ?

Discrete-time stationnary processes : What do we need them for ?

What do we mean by stationarity ?

Strict-sense stationnarity :

$X[n]$: discrete-time stochastic process

$$\forall p \in \mathbb{N}, \forall (n_1, \dots, n_p) \in \mathbb{Z}^p, \forall k \in \mathbb{Z},$$

$$\{X_{n_1}, \dots, X_{n_p}\} \stackrel{\text{law}}{=} \{X_{n_1+k}, \dots, X_{n_p+k}\}$$

Second-order (wide-sense) stationnarity :

- $X[n]$: discrete-time stochastic process
- $\text{Var}(X[n]) < +\infty$
- $\forall n, \text{E}(X[n]) = \mu$
- $\forall n, \forall k, \text{Cov}(X[n], X[n+k]) = R_X[k]$

Theorem

A wide-sense stationnary Gaussian process is strict-sense stationnary

Problem : We want to estimate a deterministic quantity y that is a function of $\{X[n]\}_n$

The **estimator** is a r. v. Y_N that is a function of $\{X[n]\}_{0 \leq n < N}$

The "quality" of the estimator is often quantified by the MSE

$$MSE(N) = E((y - Y_N)^2)$$

Example

- $y = E(X[n])$
- $Y_N = \frac{1}{N} \sum_{k=0}^{N-1} X[k]$

Problem : y is estimated using the estimator Y_N

Two important quantities

- **Bias** : $Bias_N = E(y - Y_N)$
→ (asymptotically) unbiased estimator ?
- **Variance** : $Var_N = E(Y_N^2) - E(Y_N)^2$
→ consistent estimator ?

Theorem

$$MSE(N) = Bias_N^2 + Var_N$$

An estimation example (Mean estimation):

- $y = E(X[n])$
- $Y_N = \frac{1}{N} \sum_{n=0}^{N-1} X[k]$ (ergodicity)

Let's study

- Bias ?
- Consistency ?

An estimation example (Mean estimation):

- $y = E(X[n]) = E(X)$
- $Y_N = \frac{1}{N} \sum_{n=0}^{N-1} X[k]$ (ergodicity)

Bias :

$$\begin{aligned}E(Y_N - y) &= E\left(\frac{1}{N} \sum_{n=0}^{N-1} X[k] - E(X[n])\right) \\&= \frac{1}{N} \sum_{n=0}^{N-1} E(X[k]) - E(X[n]) \\&= \frac{1}{N} \sum_{n=0}^{N-1} E(X) - E(X) \\&= 0\end{aligned}$$

This is an unbiased estimator

An estimation example (Mean estimation):

- $y = E(X[n]) = E(X)$
- $Y_N = \frac{1}{N} \sum_{n=0}^{N-1} X[k]$ (ergodicity)

Consistency (1/2) :

$$\begin{aligned}
 E(Y_N^2) - E(Y_N)^2 &= E\left(\left(\frac{1}{N} \sum_{n=0}^{N-1} X[k]\right)^2\right) - E(X)^2 \\
 &= E\left(\frac{1}{N^2} \sum_{n,k=0}^{N-1} X[k]X[n]\right) - E(X)^2 \\
 &= \frac{1}{N^2} \sum_{n,k=0}^{N-1} E(X[k]X[n]) - E(X)^2 \\
 &= \frac{1}{N^2} \sum_{n,k=0}^{N-1} R_X[n-k]
 \end{aligned}$$

An estimation example (Mean estimation):

- $y = E(X[n]) = E(X)$
- $Y_N = \frac{1}{N} \sum_{n=0}^{N-1} X[k]$ (ergodicity)

Consistency (2/2) :

$$\begin{aligned}
 E(Y_N^2) - E(Y_N)^2 &= \frac{1}{N^2} \sum_{n,k=0}^{N-1} R_X[n-k] \\
 &= \frac{1}{N} R_X[0] + \frac{2}{N^2} \sum_{l=1}^{N-1} (N - |l|) R_X[l] \\
 &= \frac{1}{N} R_X[0] + \frac{2}{N} \sum_{l=1}^{N-1} (1 - |l|/N) R_X[l]
 \end{aligned}$$

Thus if $R_X[n] \in l^1$ the operator is consistent

Another estimation example (Covariance estimation):

- $y = R_X[k]$ (we suppose $E(X) = 0$)
- (using ergodicity) $Y_N = \frac{1}{N-k} \sum_{n=0}^{N-1-k} X[n]X[n+k]$

Let's study

- Bias ?
- Consistency ?

Another estimation example (Covariance estimation):

- $y = R_X[k]$ (we suppose $E(X) = 0$)
- (using ergodicity) $Y_N = \frac{1}{N-k} \sum_{n=0}^{N-1-k} X[n]X[n+k]$

Bias :

$$\begin{aligned}
 E(Y_N - y) &= E\left(\frac{1}{N-k} \sum_{n=0}^{N-1-k} X[n]X[n+k]\right) - R_X[k] \\
 &= \frac{1}{N-k} \sum_{n=0}^{N-1-k} E(X[n]X[n+k]) - R_X[k] \\
 &= \frac{1}{N-k} \sum_{n=0}^{N-1-k} R_X[k] - R_X[k] \\
 &= 0
 \end{aligned}$$

This is an unbiased estimator

Another estimation example (Covariance estimation):

- $y = R_X[k]$ (we suppose $E(X) = 0$)
- (using ergodicity) $Y_N = \frac{1}{N-k} \sum_{n=0}^{N-1-k} X[n]X[n+k]$

Consistency ?

Another estimation example (A "better" covariance estimator):

- $y = R_X[k]$
- $Y_N = \frac{1}{N} \sum_{n=0}^{N-1-k} X[n]X[n+k]$
- Asymptotically unbiased estimator
- Consistent estimator (if $R_X[k]$ is decreasing quickly enough)

Definition : it is a **positive** bilinear operator

- $A = \sum_{i=1}^N a_i X[i]$
- $B = \sum_{i=1}^N b_i X[i]$

$$\text{Cov}(A, B) = \sum_{i,j} a_i R_X[i-j] b_j = a.(R \star b) = a.L(B)$$

⇒ The associated linear form the convolution operator

$$L(B) = R \star b$$

Problem :

**How to define the Fourier transform of a stochastic
stationnary process ?**

Definition of Power spectrum

$$\hat{R}_X(e^{i\omega}) = \sum_n R_X[n] e^{in\omega}$$

- Why is it real ?
- Why is it positive ?
- What is the inverse Fourier transform ?

An interpretation of the Power spectrum

$\hat{R}_X(e^{i\omega})$ represents the average energy contained (in average) by a realization at frequency ω

$$R_X[k] = \frac{1}{2\pi} \int_0^{2\pi} \hat{R}_X(e^{i\omega}) e^{ik\omega} d\omega$$

An important example : The white noise

- $X[n]$ second-order process
- $R_X[k] = \sigma^2 \delta[k]$
- $\hat{R}_X(e^{i\omega}) = 1$

What Estimator for the power spectrum ?

A "natural" estimator could be the **Periodogram**

$$\tilde{\hat{R}}_X(e^{i\omega}) = \left| \sum_{n=0}^{N-1} X[n] e^{-in\omega} \right|^2$$

?

A consistent estimator for the power spectrum : The averaged periodogram

$$\tilde{\hat{R}}_X(e^{i\omega}) = \frac{K}{N-1} \sum_{n=0}^{(N-1)/K-1} \left| \sum_{k=nK}^{(n+1)K-1} X[k] e^{-ik\omega} \right|^2$$

Towards the Convolution Theorem

Let $\{X[n]\}_n$ a discrete-time second order stationary process with $E(X[n]) = 0$ then

- $\forall h \in l^1$, we define $h \star X[n] = \sum_k h[n - k]X[k]$
- $E(h \star X[n]) = \sum_k h[n - k]E(X[k]) = 0$
- $\forall h \in l^1, \forall g \in l^1$, one gets

$$\forall n, n', \quad \text{Cov}(h \star X[n], g \star X[n']) = R_X \star g \star \tilde{h}[n' - n]$$

- $\forall h \in l^1$, one gets

$$\forall n, k, \quad \text{Cov}(h \star X[n], h \star X[n + k]) = R_X \star h \star \tilde{h}[k]$$

The Convolution Theorem

Let $\{X[n]\}_n$ a discrete-time second order stationary process with $E(X[n]) = 0$ then $\forall h \in l^1$, one gets

$$\hat{R}_{h*x}(e^{i\omega}) = |\hat{h}(e^{i\omega})|^2 \hat{R}_X(e^{i\omega})$$

$$\hat{R}_X(e^{i\omega}) = \sum_n R_X[n] e^{in\omega}$$

Energy formula

- Energy in the time-domain $E(X[n]^2) = R_X[0]$
- Energy in the frequency-domain

$$\frac{1}{2\pi} \int_0^{2\pi} \hat{R}_X(e^{i\omega}) d\omega$$

Energy conservation (i.e., inverse Fourier transform or $n = 0$)

$$R_X[0] = \frac{1}{2\pi} \int_0^{2\pi} \hat{R}_X(e^{i\omega}) d\omega$$

Property

- Let $h_{\omega_0, \Delta\omega}[n]$ be the band pass filter such that

$$\hat{h}_{\omega_0, \Delta\omega}|_{]-\pi, \pi[} = \sqrt{\frac{\pi}{\Delta\omega}} \left(1|_{-\omega_0 - \Delta\omega, -\omega_0 + \Delta\omega}(\omega) + 1|_{\omega_0 - \Delta\omega, \omega_0 + \Delta\omega}(\omega) \right)$$

- Let $X_{\omega_0, \Delta\omega}[n] = h_{\omega_0, \Delta\omega} * X[n]$

What is the energy of the limit process

$$\lim_{\Delta\omega \rightarrow 0} X_{\omega_0, \Delta\omega}[n] \ ?$$

$$\hat{h}_{\omega_0, \Delta\omega}|_{-\pi, \pi} = \sqrt{\frac{\pi}{\Delta\omega}} \left(1|_{-\omega_0 - \Delta\omega, -\omega_0 + \Delta\omega}(\omega) + 1|_{\omega_0 - \Delta\omega, \omega_0 + \Delta\omega}(\omega) \right)$$

Energy of $X_{\omega_0, \Delta\omega}[n]$ writes

$$\begin{aligned} R_{X_{\omega_0, \Delta\omega}}[0] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \hat{R}_{X_{\omega_0, \Delta\omega}}(e^{i\omega}) d\omega \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} |\hat{h}_{\omega_0, \Delta\omega}(e^{i\omega})|^2 \hat{R}_X(e^{i\omega}) d\omega \\ &= \frac{1}{2\Delta\omega} \left(\int_{-\omega_0 - \Delta\omega}^{-\omega_0 + \Delta\omega} \hat{R}_X(e^{i\omega}) d\omega + \int_{\omega_0 - \Delta\omega}^{\omega_0 + \Delta\omega} \hat{R}_X(e^{i\omega}) d\omega \right) \\ &\longrightarrow \hat{R}_X(e^{-i\omega_0}) + \hat{R}_X(e^{i\omega_0}) \end{aligned}$$